



Solar Powered Agriculture Pump Using Dual Output Buck Boost Converter

G. PRAKASH¹, R. AZHAGUMURUGAN², K DHINESH KUMAR³, G. RAMYA⁴

¹Department of Electrical and Electronics Engineering, Sri Sai Ram Institute of Technology, Chennai, India;

² Department of Electrical and Electronics Engineering, Sri Sai Ram Engineering College, Chennai, India;

³Department of Electrical and Electronics Engineering, Narsimha Reddy Engineering College, Telangana;

⁴Department of Mechanical Engineering, Rajalakshmi Engineering College, Chennai, India;

ABSTRACT- Photovoltaic (PV) framework is proposed for rustic applications like water siphoning for water system, day by day needs and drinking water. The dire need of power was fulfilled by this framework. Battery is utilized to store the power creating from the sunlight based charger .assuming we run the heap Dc-Dc converter is compulsory .This framework proposes With the expanding request of utilizations that have two diverse result voltages, the single-input double result (SIDO) converter with less parts is turning into the savvy choice as opposed to utilizing two single information single result converters. Attributable to the utilization of a voltage multiplier circuit, a high move forward voltage transformation proportion is accomplished with generally low voltage stress of switches.

I INTRODUCTION

DC/DC power converters with numerous results are broadly utilized in applications like mixture electric vehicles, DC miniature framework frameworks, sustainable power frameworks, and PCs that require energy stream and voltage guideline of various results. A straightforward method for understanding these prerequisites is utilizing N single-input single-yield (SISO) as outlined which brings about significant expense and low power effectiveness because of an enormous number of parts. To lessen the quantity of parts and in general expense, the single-input multi-yield (SIMO) converter as displayed turns into the predominant choice since it can incorporate numerous SISO converters into one.

In extra auxiliary windings or transformers are used in regular disengaged SISO converters to acquire numerous results, whose cost is low attributable to the basic design. There is just one control variable, and the other

result voltages are directed relying upon the turns proportions of their relating windings or transformers. Nonetheless, the heap variety of one result port will cause overshoot or undershoot voltages of different results because of parasitic attributes of parts, which is called cross guideline issue. Henceforth, voltage blunder of the unregulated result ports is critical. To accomplish free control of each result, optional side post-controllers (SSPRs) are introduced to supply extra control factors for each result by adding outer dynamic switches in the auxiliary circuits. Nonetheless, the cross-guideline issue actually exists during the powerful reaction with load variety in light of the fact that the flows of all heaps course through a similar polarized inductor. For SSPRs arrangements, single-inductor multi-yield converters are the promising geographies since they utilize just a single inductor and few switches, in this manner the expense is low. Sadly, it additionally experiences the cross-guideline issue in the consistent conduction mode (CCM) because of the interface among the heaps. Different medicinal strategies are proposed to stifle the cross guideline of single-inductor multi-yield converters the time multiplexing control method dependent on irregular conduction mode is utilized to work on the cross-guideline execution. In any case, the inductor current wave is enormous in the weighty burden condition, bringing about a huge result voltage wave and exchanging commotion.

Moreover, an extra switch is associated with the inductor in corresponding to acknowledge pseudo proceeds with conduction mode (PCCM), which lessens the high current wave of parts while applying weighty burdens. Notwithstanding, the power misfortune is expanded because of the presence of nonzero inductor current during the freewheeling stretch.

II Z SOURCE CONVERTER

Z source inverter is a solitary stage power converter and it is gotten from the impedance source inverter geography displayed in fig.1. The conventional voltage source and current source inverters experiences the restriction of setting off two switches in a similar leg or stage prompts a source cut off the greatest possible result voltage can't surpasses the DC include and can deliver a voltage lower than DC input voltage. z-source inverter defeat these downsides by using the few shoot through zero states. Shoot through zero state is created when every one of the upper switches or lower switches are terminated at the same time to support up the result voltage.

The impedance network couples the source and the inverter to accomplish voltage help in a solitary stage. Various geographies of z source inverter are Quasi-z source inverter, semi-Z-source inverter, Trans-Z source inverter, Y-source inverter, X-source inverter, and so on

Z-source and its inferred geographies could be applied in numerous application regions like energy component, photovoltaic and wind power age framework, half breed electric vehicle , etc.

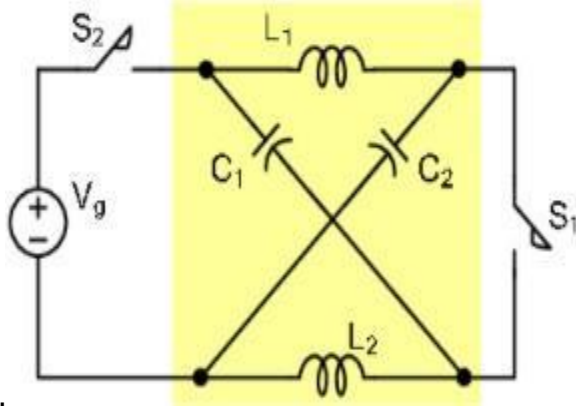


Fig.1.Z source converter

The DC source can be taken as a battery, diode rectifier, thyristor converter or PV cluster. Here, diode rectifier is utilized for the DC source. The info AC supply is changed over to DC by utilizing the diode span rectifier. The circuit includes impedance network having inductances L_1 , L_2 and capacitances C_1 , C_2 and the inverter module contains IGBT switches with an enemy of equal diode. Voltage of capacitor C_4 is close to steady DC voltage, since the AC input voltage is amended by the diode span rectifier and separated by the capacitors C_3 and C_4 .

The controllable gadget VT_5 is utilized to manage the spasmodic activity condition when the heap is light, it very well may be an IGBT paddle a Power MOSFET. Here VT_5 is picked as IGBT gadget. IGBT part is turned ON when the inverter works in non-shoot through state for example in the dynamic state and conventional zero state. IGBT part is wound down when inverter works in shoot through zero state.

what's more they could likewise be applied in ac power supply because of their buck-support work. The primary circuit utilizes voltage-took care of Z source inverter structure. This framework is reasonable to wide scope of burden, from no-heap, light burden to substantial burden.

ZSI circuit varies from that of regular ZSI in the LC impedance network interface among source and inverter. This organization shields the circuit from harm when shoot through happens and by utilizing the shoot through state, ZSI network helps the DC interface voltage.

III PROPOSED SYSTEM

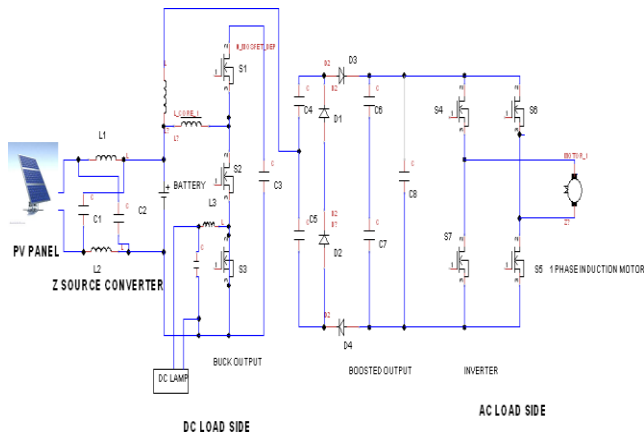


Fig.2 Proposed Circuit Diagram

The circuit chart contains diodes front end, channel capacitors, impedance organization, inverter switches and burden is displayed in fig.2. When the Buck and the lift converters are associated in antiparallel across one another with the subsequent circuit is essentially having a similar design as the basic Boost and Buck structure however with the additional component of bidirectional power stream. The underneath figure shows the essential design of the Non-Isolated Half-Bridge Bidirectional DC converter displayed in fig.3.

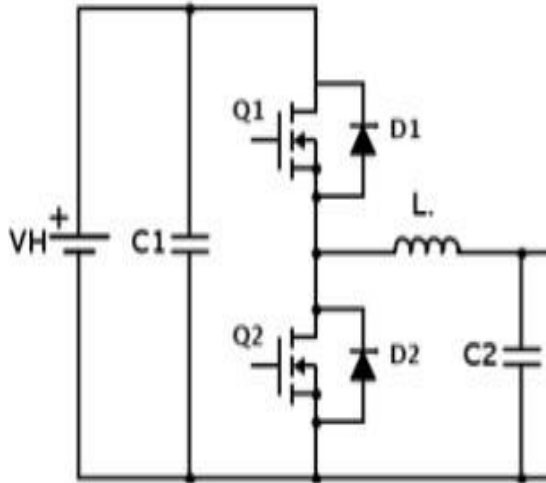


Fig.3. Bidirectional DC-DC Converter

Non-Isolated Half-Bridge Bidirectional DC Converter coming about out of the counter equal association of the Buck and Boost converters

The above circuit can be made to work in buck or lift mode relying upon the exchanging of

the Mosfets Q1 and Q2. The switches Q1 or Q2 in blend with the counter equal diodes D1 or D2 (going about as freewheeling diode) separately, makes the circuit move forward or venture down the voltage applied across them. The bidirectional activity of the above circuit can be clarified in the under two modes as follows:

Mode 1 (Boost Mode): In this mode switch Q2 and diode D1 goes into conduction relying upon the obligation cycle while the switch Q1 and diode D2 are off constantly. This mode can additionally be separated into two spans relying upon the conduction on the switch Q1 and diode D2 as displayed in the Fig.3

Span 1 (Q2-on, D2-off ; Q1-off, D2-Off): In this mode Q2 is on and consequently can be viewed as shortcircuited, hence the lower voltage battery charges the inductor and the inductor current continues expanding till not the door beat is taken out from the Q2. Likewise, since the diode D1 is turned around one-sided in this mode and the switch Q1 is off, no current moves through the switch Q1.Interval 2 (Q1-off, D1-off; Q2-off, D2-on): In this mode Q2 and Q1 both are off and consequently can be viewed as opened circuited. Presently since the current coursing through the inductor can't change immediately, the extremity of the voltage across it switches and henceforth it begins acting in series with the info voltage. Accordingly, the diode D1 is forward one-sided and henceforth the inductor current charges the result capacitor C2 to a higher voltage. Consequently, the result voltage helps up.

Mode 2 (Buck Mode): In this mode switch Q1 and diode D2 goes into conduction relying upon the obligation cycle though the switch Q2 and diode D1 are off constantly. This mode can additionally be isolated into two stretches relying upon the conduction on the switch Q2 and diode D1 as displayed in the Fig.

Span 1 (Q2-on, D2-off; Q1-off, D2-Off): In this mode Q1 is on and Q2 is off and consequently the same circuit is as displayed in the Fig beneath. The higher voltage battery will charge the inductor and the result capacitor will get charged by it.

Span 2 (Q1-off, D1-off; Q2-off, D2-on): In this mode Q2 and Q1 both are off. Again since the inductor current can't change quickly, it gets released through the freewheeling diode D2. The voltage across the heap is ventured down when contrasted with the info voltage.

An examination between the various elements of the non-confined bidirectional geographies have been introduced underneath:

1. During move forward mode, in the buck-help bidirectional converter the rms worth of the current through the inductor and the power switches is more prominent by a sum equivalent to the result current when contrasted with the buck-support course bidirectional converter. Additionally, the capacitor rms current likewise surpasses in the previous case by a measure of the 1/third of the result current. Accordingly, in the bidirectional buck-

support converter the inductor, power switches and the capacitor work under more warm and electrical anxieties when contrasted with the buck help course converter bringing about the more prominent power misfortune and furthermore causing the immersion of the inductor center. Additionally, since the weight on the mosfet and the diode is higher, buck-support bidirectional converter requires power gadgets with bigger gadget appraisals. Higher rms flows additionally bring about higher conduction misfortunes and along these lines diminishing the general proficiency of the buck-support bidirectional converter.

2. However, the quantity of gadgets needed by the course buck help converter is double the number gadgets in buck-support bidirectional converter. This issue can be tackled by utilizing the Half-Bridge Bidirectional DC Converter. It has similar no gadgets as the buck-help bidirectional converter and can be utilized rather than the buck-support course bidirectional converter for the applications that require the lift activity just one way and the buck in the other.

3. The principle benefits of the half extension bidirectional converter when contrasted with the bidirectional Cuk converter is that it just requires one inductor rather than two and that too a large portion of the worth of last option just as the power switches evaluations needed for the half scaffold bidirectional converter is a lot of lower when contrasted with the Cuk converter. Likewise, the productivity of the half scaffold converter is higher than the Cuk converter due to the lower inductor current and along these lines lower conduction just as lower exchanging misfortunes.

IV. PROPORTIONAL-INTEGRAL CONTROLLER

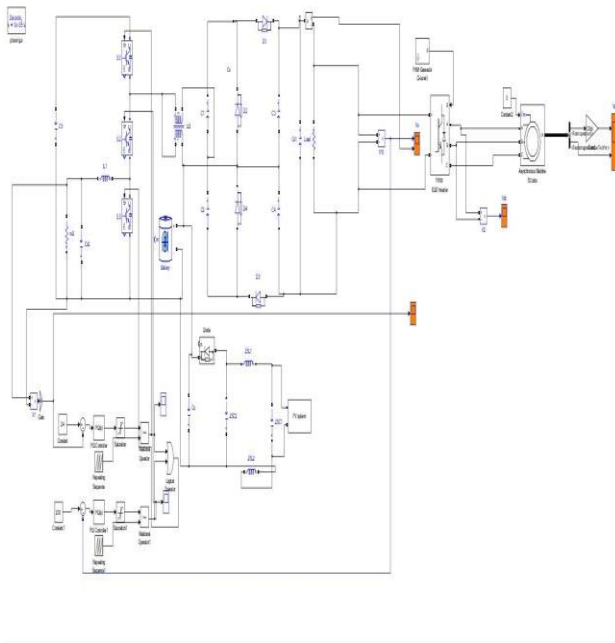
A variety of Proportional Integral Derivative (PID) control is to utilize just the relative and vital terms as PI control. The PI regulator is the most famous variety, considerably more than full PID regulators. The worth of the regulator yield $u(t)u(t)$ is taken care of into the situation as the controlled variable input. $e(t)=SP-PV$

The ubias term is a constant that is typically set to the value of $u(t)$ when the controller is first switched from manual to automatic mode. This gives "bumpless" transfer if the error is zero when the controller is turned on. The two tuning values for a PI controller are the controller gain, K_c and the integral time constant τ_I . The value of K_c is a multiplier on the proportional error and integral term and a higher value makes the controller more aggressive at responding to errors away from the set point. The set point (SP) is the target value and process variable (PV) is the measured value that may deviate from the desired value. The error from the set point is the difference between the SP and PV and is defined as $e(t)=SP-PV$.

Term P is relative to the current worth of the SP-PV blunder $e(t)$. For instance, in the event

that the blunder is enormous and positive, the control result will be proportionately huge and positive, considering the addition factor "K". Utilizing corresponding control alone will forever bring about a blunder between the setpoint and the genuine cycle esteem, since it requires a mistake to produce the relative reaction. Assuming there is no blunder, there is no restorative reaction.

Term I represents past upsides of the SP-PV mistake and incorporates them after some time to deliver the I expression. For instance, assuming there is a leftover SP-PV blunder later the use of corresponding control, the basic term looks to dispense with the lingering mistake by adding a control impact because of the noteworthy total worth of the blunder. At the point when the blunder is killed, the basic term will stop to develop. This will bring about the relative impact lessening as the blunder diminishes, yet this is made up for by the developing basic impact.



The balance of these effects is achieved by "loop tuning" to produce the optimal control function. The tuning constants are displayed beneath as "K" and should be determined for each Cntrol application, as they rely upon the reaction attributes of the total circle outside to the regulator. sensor, the last control component (like a control valve), any control signal deferrals and the actual interaction. Inexact upsides of constants can typically be at first entered knowing the kind of utilization, however they are regularly refined, or tuned, by "knocking" the interaction practically speaking by, for example, presenting a setpoint change and noticing the framework reaction

The numerical model and reasonable circle above both utilize a "immediate" control activity

5075 | G. PRAKASH **Solar Powered Agriculture Pump Using Dual Output Buck Boost Converter**

for every one of the terms, which implies an expanding positive blunder brings about an expanding positive control yield for the added terms to apply revision. Nonetheless, the result is classified "turn around" acting assuming it's important to apply negative remedial activity. For example, assuming the valve in the stream circle was 100-0% valve opening for 0-100% control output meaning the regulator activity must be turned around. Some cycle control plans and last control components require this converse activity. A model would be a valve for cooling water, where the fizzle be a valve for cooling water, where the fail-safe mode, in the case of loss of signal, would be 100% opening of the valve; therefore 0% controller output needs to cause 100% valve opening.

V. SIMULATED CIRCUIT DIAGRAM

The fig.4 shows the simulated circuit diagram. The output of Buck and Boost waveform shown in fig. 5 and fig. 6. The motor speed and torque output is shown in fig.7.

Fig.4 Simulated Circuit Diagram

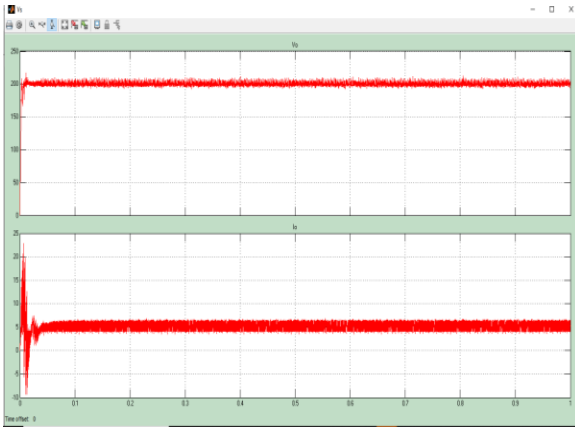


Fig.5 Boosted Output Waveform

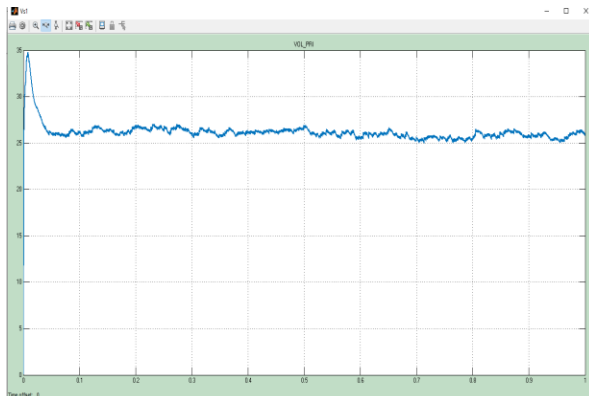


Fig. 6 Bucked Output Waveform

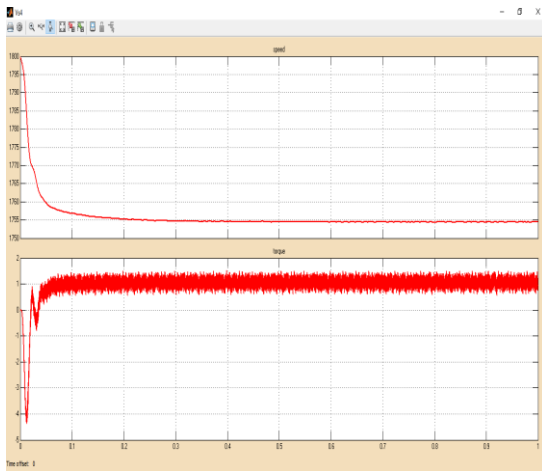


Fig.7 Motor Speed and Torque Curve

VI. CONCLUSION AND FUTURE SCOPE

This framework presents a coordinated SIDO converter comprising of a buck converter and a voltage multiplier circuit, which enjoys benefits as concurrent buck and lift yield voltages and high-power effectiveness. The activity examination, plan thought, explore check, and correlation are talked about exhaustively. By utilizing SIDO converter, the proposed geography utilized for Areas and farming area.in SIDO converter work as autonomous move forward and - down yield voltages simultaneously. Z source converter is utilized to keep up with the pv power at shady condition. enlistment engine is associated through inverter at help yield side .in other hand buck yield side we are run dc light

Nowadays researches going on Renewable energy resources and how it will be interfaced with converters.in this project am using only solar panel, in cloudy condition am replacing or adding small wind turbine as addition source for this system.

REFERENCES

- [1] K.W.Lee, D.K. Kim, B.T.kim and B.I. Kwon, "A novel staring method of the surface permanent- magnet BLDC motors without position sensor for reciprocating compressor," IEEE Trans. Ind Appl., vol. 44 no 1, pp. 85-92, Jan/Feb. 2008.
- [2] D.K. Kim, K.W. Lee and B.I Kwon, "Commutation torque ripple reduction in a position sensorless brushless DC motor drive," IEEE Tran Power Electron, vol.21, no. 6, pp. 1762-1768, Nov 2006.

- [3] F. Rodriguez and A. Emadi, "A novel digital control technique for brushless DC motor drives," *IEEE Trans. Ind. Electron.*, vol. 54, no. 5, pp. 2365-2373, Oct. 2007.
- [4] Caricchi.F, GiulliiCapponi.F, Crescimbin.F and Solero.L "Sinusoidal Brushless Drive with Low-Cost Linear Hall Effect Position Sensors" *IEEE Conf.* pp.799-804, 2001.
- [5] Chun-Liang Lin, Horn-Yong Jan, and Niahn-Chung Shieh "GA-Based Multi objective PID Control for a Linear Brushless DC Motor" *IEEE Trans. MECH*, Vol. 8, No. 1, pp. 56 – 65, March 2003.
- [6] Gui-Jia Su, and John W. McKeever "Low-Cost Sensorless Control of Brushless DC Motors with Improved Speed Range" *IEEE Trans. Pow. Electron*, Vol. 19, no. 2, pp. 296–302, March 2004. [5] Hang C-C. K. J. Astrom and W. K. Ho, "Refinements of the ZieglerNichols tuning formula", *IEEE Proc. Part D*, vol. 138, No. 2, pp. 11 1-1 18, March 1991.
- [7] J.-H. Lee, T.-S.Kim, and D.-S. Hyun, "A study for improved of speed response characteristic in four-switch three-phase BLDC motor," in *Proc. IEEE Ind. Electron. Soc. Conf.*, 2004, vol. 2, pp. 1339–1343.
- [8] P. Pillay and R. Krishnan, "Modeling, simulation and analysis of permanent- magnet motor drives. II. The brushless DC motor drive," *IEEE Trans. Ind. Appl.*, vol. 25, no. 2, pp. 274–279, Mar./Apr. 1989
- [9] A. K. Mishra and B. Singh, "A PV fed DC-DC converter for switched reluctance motor driven agriculture pump," in *Proc. National Power Electronics Conference (NPEC)*, Pune, 2017, pp. 12-18.
- [10] V. Indragandhi, R. Selvamathi and T. Arunkumari, "Speed control of a switched reluctance motor using PID controller for PV based water pumping applications," in *Proc. Innovations in Power and Advanced Computing Technologies (i-PACT)*, Vellore, 2017, pp. 1-4.
- [11] X. Wang, C. G., Y. H., W. Cao and X. Chen, "Renewable energy-fed switched reluctance motor for PV pump applications," in *Proc. of IEEE Conf. and Expo, Transportation Elect. Asia-Pacific (ITEC AsiaPacific)*, 2014, Beijing, 2014.
- [12] V. Narayana, A. K. Mishra and B. Singh, "Design of SRM driven BESS based PV powered water pumping system," in *Proc. 7th Power India International Conference (PIICON)*, Bikaner, 2016, pp. 1-6.
- [13] B. Singh, A. K. Mishra and R. Kumar, "Solar Powered Water Pumping System Employing Switched Reluctance Motor Drive," *IEEE Transactions on Industry Applications*, vol. 52, no. 5, pp. 3949- 3957, Sept.-Oct. 2016.
- [14]] M. H. Uddin, M. A. Baig and M. Ali, "Comparision of 'perturb & observe' and 'incremental conductance', maximum power point tracking algorithms on real

environmental conditions,” in Proc. International Conference on Computing, Electronic and Electrical Engineering (ICE Cube), Quetta, 2016, pp. 313-317.

- [15] G. Yüksek and A. N. Mete, “A hybrid variable step size MPPT method based on P&O and INC methods,” in Proc. 10th International Conference on Electrical and Electronics Engineering (ELECO), Bursa, 2017, pp. 949-953.